

# A Bayesian Approach of Hyperanalytic Denoising

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**Abstract** – The property of shift-invariance associated with a good directional selectivity are important for the application of a wavelet transform, (WT), in many fields of image processing. Generally, complex wavelet transforms, like for example the Double Tree Complex Wavelet Transform, (DTCWT), poses these good properties. In this paper we propose the use of a new implementation of such a WT, recently introduced, namely the hyperanalytic wavelet transform, (HWT), in denoising applications. The proposed denoising method is very simple, implying three steps: the computation of the forward WT, the filtering in the wavelets domain and the computation of the inverse WT, (IWT). The goal of this paper is the association of a new implementation of the HWT, recently proposed, with a maximum a posteriori (MAP) filter. Some simulation examples and comparisons prove the performances of the proposed denoising method.

**Keywords** – Directional selectivity, Hyperanalytic wavelet transform, Image denoising, Maximum a posteriori filter.

## I. INTRODUCTION

During acquisition and transmission, images are often corrupted by additive noise that can be modeled as Gaussian most of the time. The aim of an image-denoising algorithm is then to reduce the noise level, while preserving the image features. Such a system must realize a great noise reduction in the homogeneous regions and the preservation of the details of the scene in the other regions. There is a great diversity of estimators used like denoising systems. A possible classification criterion for these systems takes into account the theory that is found at the basis of each one. In this respect, there are two categories of denoising systems, those based on wavelet theory and the others. In fact, David Donoho, introduced the word denoising in association with the wavelet theory, [1]. From the second category, taking into account their performance, we must mention the denoising systems proposed in [2] and [3]. The denoising system proposed in [2] is based on the shape-adaptive DCT (SA-DCT) transform that can be computed on a support of

arbitrary shape. The SA-DCT is used in conjunction with the Anisotropic Local Polynomial Approximation (LPA) - Intersection of Confidence Intervals (ICI) technique, which defines the shape of the transform support in a pointwise adaptive manner. Since supports corresponding to different points are in general overlapping, the local estimates are averaged together using adaptive weights that depend on the region's statistics. The denoising system proposed in [3] is a MAP filter that acts in the spatial domain. It makes a different treatment of regions with different homogeneity degree. These regions can be treated independent with the same MAP filter choosing between different prior models. The multi-resolution analysis performed by the WT has been shown to be a powerful tool to achieve good denoising. In the wavelet domain, the noise is uniformly spread throughout the coefficients, while most of the image information is concentrated in the few largest ones (sparsity of the wavelet representation), [4-7]. The corresponding denoising methods have three steps, [1]: 1) The computation of the forward WT, 2) the filtering of the wavelet coefficients, 3) the computation of the IWT of the result obtained. Numerous WTs can be used to operate these treatments. The first one was the Discrete Wavelet Transform, DWT, [1]. It has three main disadvantages, [8]: lack of shift invariance, lack of symmetry of the mother wavelets and poor directional selectivity. These disadvantages can be diminished using a complex wavelet transform [8, 9]. Over twenty year ago, Grossman and Morlet [10] developed the Continuous Wavelet Transform (CWT) [11], using continuous complex-valued mother wavelets. Initial analysis based on wavelet decompositions was implemented using such mother wavelets. Both magnitude and phase descriptions of non-stationary signals were determined, and an early example of analysis includes wavelet ridge methods proposed by Delprat *et al.* [12]. However subsequently for many years interest focused on the Discrete Wavelet Transform (DWT) and signal estimation. The DWT was developed to implement the WT of time-compact mother wavelets and as compact discrete wavelet filters cannot be exactly analytic [13], real wavelets were used. A revival of interest in later years has occurred in both

signal processing and statistics for the usage of complex wavelets, [14], and in particular complex analytic wavelets [15]–[17]. This revival of interest may be linked to the development of complex-valued discrete wavelet filters [18] and the clever dual filter bank [15, 19]. The complex wavelet transform has been shown to provide a powerful tool in signal and image analysis [11], where most of the properties of the transform follow from the analyticity of the wavelet function. In [20] were derived large classes of wavelets generalizing the concept of a 1-D local complex-valued analytic decomposition to a 2-D vector-valued hyperanalytic decomposition. In the present paper we propose the utilization of a very simple implementation of the HWT, recently proposed, [21]. It has a high shift-invariance degree versus other quasi-shift-invariant WTs at same redundancy. It has also an enhanced directional selectivity. All the WTs have two parameters: the mother wavelets, MW and the primary resolution, PR, (number of iterations). The importance of their selection is highlighted in [23]. Another appealing particularity of those transforms, becoming from their multiresolution capability, is the interscale dependency of the wavelet coefficients. Numerous non-linear filter types can be used in the WT domain. A possible classification is based on the nature of the useful component of the image to be processed. Basically, there are two categories of filters: those built supposing that the useful component of the input image is deterministic and those based on the hypothesis that this component is random. To the first category belong the hard-thresholding filter, [1], the soft-thresholding filter, [1,11], that minimizes the Min-Max estimation error and the Efficient SURE-Based Inter-scales Pointwise Thresholding filter [24], that minimizes the Mean Square Error, (MSE). Filters obtained by minimizing a Bayesian risk, typically under a quadratic cost function (a delta cost function (MAP estimation [4,5,7]) or the minimum mean squared error - MMSE estimation [6]) belong to the second category. The construction of MAP filters supposes the existence of two statistical models, for the useful component of the input image and for its noise component. The MAP estimation of  $u$ , realized using the observation  $y=u+n$ , (where  $n$  represents the WT of the noise and  $u$  the WT of the useful component of the input image) is given by the following relation, called MAP filter equation:

$$\hat{u}(y) = \underset{u}{\arg \max} \{ \ln(f_n(y-u)f_u(u)) \} \quad (1)$$

where  $f_x$  represents the probability density function (pdf) of  $x$ . Generally, the noise component is supposed Gaussian distributed. For the useful component there are many models. This distribution changes from scale to scale. For the first iterations of the WT it is a heavy tailed distribution and with the increasing of iterations number it converges to a Gaussian. There are two solutions to deal with this mobility. The first one supposes to use a fixed simple model, risking an increasing of imprecision across the scales. This way, there is a chance to obtain a closed form input-output relation for the MAP filter. This is the case of the bishrink filter [5]. An explicit input-output relation has two advantages: it simplifies

the implementation of the filter and it permits the analysis of its sensitivities. The second solution supposes to use a generalized model, defining a family of distributions and the identification of the best fitting element of this family for the distribution of the wavelet coefficients at a given scale. For example in [4] is used the family of Pearson's distributions, in [7] the family of SaS distributions and in [25] is used the model of Gauss–Markov random field. The use of such generalized model makes the treatment more precise but implies implicit solutions for the MAP filter equation, it can be solved only numerically and the sensitivities of the filter obtained cannot be evaluated. If the pdfs  $f_w$  and  $f_n$  do not take into account the interscale dependency of the wavelet coefficients than the MAP filter obtained is called marginal. This paper proposes a new denoising method for images based on the association of the new implementation of HWT, already mentioned, [21], with a very simple marginal MAP filter. The second section is dedicated to the new implementation of the HWT. The third section presents this marginal MAP filter. The aim of the forth section is the presentation of some simulation results. The paper concludes with few final remarks.

## II. A NEW IMPLEMENTATION OF THE HWT

All the WTs already mentioned have simpler or more complicated 2D generalizations. The generalization of the analyticity concept in 2D is not obvious, because there are multiple definitions of the Hilbert transform in this case. In the following we will use the definition of the analytic signal associated to a 2D real signal named hypercomplex signal. So, the hypercomplex mother wavelets associated to the real mother wavelets  $\psi(x, y)$  is defined as:

$$\begin{aligned} \psi_a(x, y) = & \psi(x, y) + i\mathcal{H}_x\{\psi(x, y)\} + \\ & + j\mathcal{H}_y\{\psi(x, y)\} + k\mathcal{H}_x\{\mathcal{H}_y\{\psi(x, y)\}\} \end{aligned} \quad (2)$$

where  $i^2 = j^2 = -k^2 = -1$ , and  $ij = ji = k$ , [22].

The HWT of the image  $f(x, y)$  is:

$$HWT\{f(x, y)\} = \langle f(x, y), \psi_a(x, y) \rangle. \quad (3)$$

Tacking into account relation (2) it can be written:

$$\begin{aligned} HWT\{f(x, y)\} = & DWT\{f(x, y)\} + \\ & iDWT\{\mathcal{H}_x\{f(x, y)\}\} + jDWT\{\mathcal{H}_y\{f(x, y)\}\} + \\ & + kDWT\{\mathcal{H}_y\{\mathcal{H}_x\{f(x, y)\}\}\} = \\ & \langle f_a(x, y), \psi(x, y) \rangle = DWT\{f_a(x, y)\}. \end{aligned} \quad (4)$$

So, the 2D-HWT of the image  $f(x, y)$  can be computed with the aid of the 2D-DWT of its associated hypercomplex image. The new HWT implementation, [21], presented in figure 1, uses four trees, each one implementing a 2D-DWT. The first tree is applied to the input image. The second and

the third trees are applied to 1D discrete Hilbert transforms computed across the lines ( $\mathcal{H}_x$ ) or columns ( $\mathcal{H}_y$ ) of the input image. The fourth tree is applied to the result obtained after the computation of the two 1D discrete Hilbert transforms of the input image. The enhancement of the directional selectivity of the 2D-HWT is realized like in the case of the 2D-DTCWT, [9,19], by linear combinations of detail coefficients belonging to each subband of each of the four 2D DWTs. Let us consider, for example, the case of the diagonal detail subbands, (HH), presented in figure 2. We selected a particular input image,  $f(x,y) = \delta(x,y)$ , to appreciate the frequency responses associated to different transfer functions represented in figure 1. More precisely, the example in figure 2 refers to the transfer functions that relate the input  $f$  with the outputs  $z_{-r}$  and  $z_{+r}$ . The spectrum of the input image,  $\mathcal{F}\{\delta(x,y)\}(f_x, f_y)$  is constant. The wavelet coefficients belonging to the subband HH are obtained by lines and columns high-pass filtering. We have supposed ideal high-pass filters. The spectra of the wavelet coefficients  $d_1, d_2, d_3, d_4$  belonging to the subband HH,  $\mathcal{F}\{DWT_{HH}\{\delta(x,y)\}\}$ ,  $\mathcal{F}\{DWT_{HH}\{\mathcal{H}_x\{\delta(x,y)\}\}\}$ ,  $\mathcal{F}\{DWT_{HH}\{\mathcal{H}_y\{\delta\}\}\}$ ,  $\mathcal{F}\{DWT_{HH}\{\mathcal{H}_y\{\mathcal{H}_x\{\delta\}\}\}\}$ , have two preferential orientations, corresponding to the two diagonals ( $\pm\pi/4$ ). So, the 2D DWT cannot separate these two orientations. But the spectra of the coefficients obtained after linear combinations, for example  $z_{-r}$  and  $z_{+r}$ ,  $\mathcal{F}\{^{HH}z_{-r}\}(f_x, f_y)$  and  $\mathcal{F}\{^{HH}z_{+r}\}(f_x, f_y)$  have only one

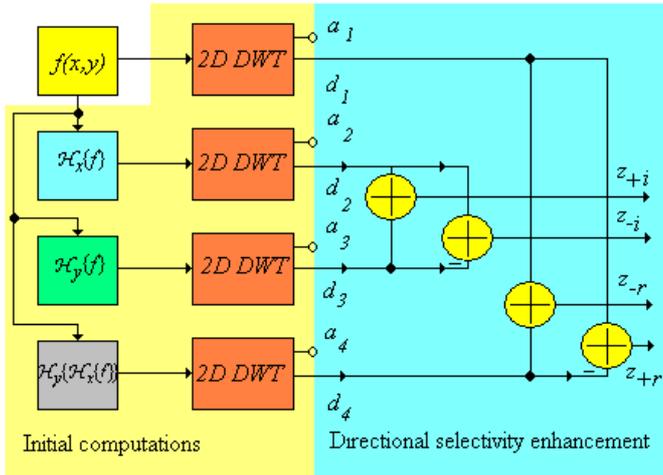


Figure 1. The new 2D-HWT-implementation architecture.

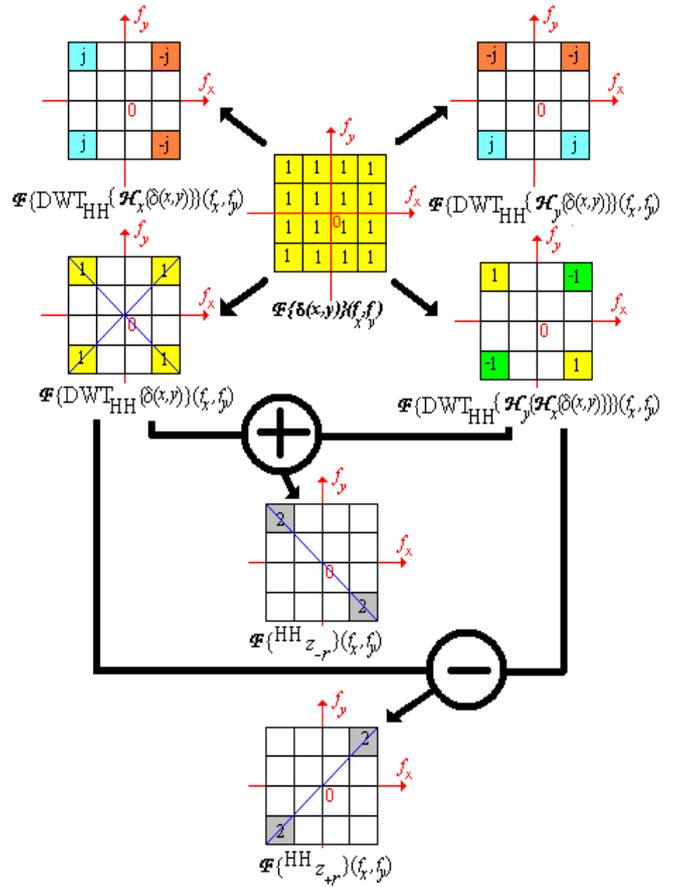


Figure 2. The strategy of directional selectivity enhancement in the HH subband. The frequency responses of the systems that transform the input image  $f$  into the output diagonal detail coefficient sub-images  $z_{-r}$  and  $z_{+r}$  represented in figure 1.

preferential direction, the second diagonal respectively the first one. So, using the 2D HWT these directions can be separated. The same strategy can be used to enhance the directional selectivity in the other two subbands: LH and HL, obtaining the preferential orientations:  $\pm\text{atan}(2)$  and  $\pm\text{atan}(1/2)$ . A comparison of the directional selectivity of the 2D DWT and the proposed implementation of the 2D HWT is presented in figure 3. We have conceived a special input image, in the frequency domain, to conduct this simulation. Its spectrum, represented in figure 3, is oriented following the directions:  $0, \pm\text{atan}(1/2), \pm\pi/4, \pm\text{atan}(2)$  and  $\pi$ . The better directional selectivity of the new implementation of 2D-HWT versus the 2D-DWT can be easily observed. For example, the new implementation makes the difference between the two principal diagonals or between the directions  $\pm\text{atan}(1/2)$  whereas the DWT cannot make such differences.

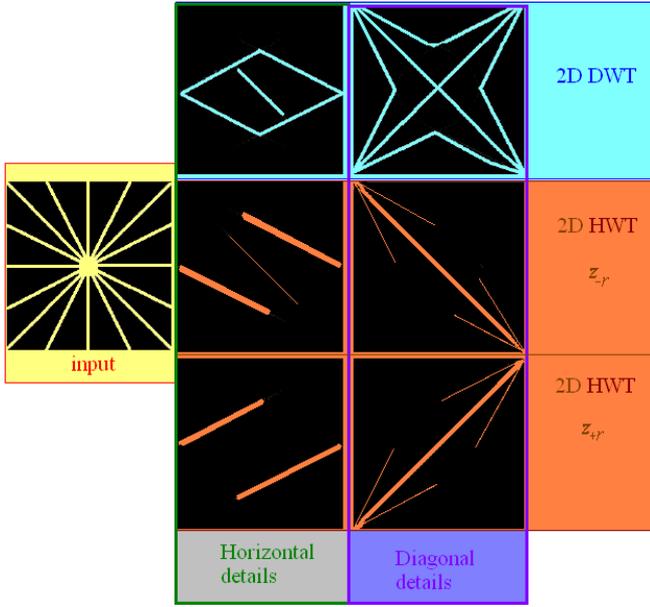


Figure 3. The absolute values of the spectra of horizontal and diagonal detail sub-images obtained after the first iterations of 2D DWT and 2D HWT (proposed implementation). In the HWT case, the directional selectivity was enhanced using linear combinations.

### III. THE MARGINAL MAP FILTER

In the following, we will consider a Gaussian distribution for the noise coefficients ( $f_n$ ) and a Laplacian distribution for the useful signal coefficients ( $f_u$ ). The noise coefficients have zero mean and variance  $\sigma_n^2$ . Regarding the a-priori statistical assumptions made, note that the wavelet transform of an image consists into a small number of high value wavelet coefficients (especially marking the contours) and a large number of small value coefficients (for the homogeneous regions). A heavy-tailed distribution for these coefficients seems therefore far more realistic than a Gaussian-one, and the particular case of a Laplacian pdf becomes attractive by its computational tractability.

#### A. The solution of the MAP filter equation

Consequently, we take:

$$f_u(u) = \frac{1}{\sqrt{2}\sigma_u} \exp\left(-\frac{\sqrt{2}|u|}{\sigma_u}\right) \quad (5)$$

Under the considered hypothesis, the equation (1) becomes:

$$\frac{y-\hat{u}}{\sigma_n^2} - \frac{\sqrt{2}}{\sigma_u} \text{sgn } \hat{u} = 0. \quad (6)$$

Finally, the solution can be expressed as:

$$\hat{u} = \text{sgn}(y) \left( |y| - \sqrt{2}\sigma_n^2/\sigma_u \right)_+, \quad (7)$$

where  $(X)_+ = X$  for  $X > 0$  and 0 otherwise. In the equation (7), we denoted by  $\sigma_n^2$  the noise variance and by  $\sigma_u$  the standard

deviation of the useful image coefficients. The relation (7) reduces to a soft-thresholding of the noisy coefficients with a variable threshold. In practice, the statistical parameters in (7) are not known and therefore they must be estimated. To estimate the noise variance  $\sigma_n^2$  from the noisy wavelet coefficients, a robust median estimator is used from the finest scale wavelet coefficients corresponding to each of the for DWTs:

$$\hat{\sigma}_n^2 = \frac{\text{median}(|y_i|)}{0.6745}, \quad y_i \in \text{subband HH}. \quad (8)$$

In [5], the marginal variance of the  $k$ 'th coefficient is estimated using neighboring coefficients in the region  $N(k)$ , a window centered at the  $k$ 'th coefficient. To make this estimation one gets  $\sigma_y^2 = \sigma_u^2 + \sigma_n^2$ , where  $\sigma_y^2$  is the marginal variance of noisy observations,  $y$ . For the estimation of  $\sigma_y^2$ , is proposed the following relation:

$$\hat{\sigma}_y^2 = \frac{1}{M} \sum_{y_i \in N(k)} y_i^2, \quad (9)$$

where  $M$  is the size of the neighborhood  $N(k)$ . Then  $\sigma_u$  can be estimated as:

$$\hat{\sigma}_u = \sqrt{\left( \hat{\sigma}_y^2 - \hat{\sigma}_n^2 \right)_+}. \quad (10)$$

#### B. Directional windows in wavelets domain

In [5] the regions  $N(k)$  were rectangular with size  $7 \times 7$ . In [26] were proposed directional windows. The algorithm in [5] uses the same squared window for all the three oriented subbands in each level of a 2D DWT, which, in fact, imposes an assumption that the energy distribution of the image in each oriented subband is isotropic. However, this is not true for most images. The energy clusters in the horizontal, vertical, and diagonal subbands are mainly distributed along the horizontal, vertical, and diagonal directions, respectively. For this reason, the estimator using a squared window often leads to downward-biased estimates within and around energy clusters, which is disadvantageous to preserve the edges and texture in images. In [26], the elliptic directional windows are introduced to estimate the signal variances in each oriented subband. We generalized here this idea for the proposed implementation of the 2D HWT, using constant array elliptic estimation windows having the principal axe oriented following the directions:  $\pm \text{atan}(1/2)$ ,  $\pm \pi/4$ , and  $\pm \text{atan}(2)$ .

### IV. SIMULATION RESULTS

In the following, some simulation results, obtained using the image Lena (size  $512 \times 512$ ) additively perturbed with white Gaussian noise, are reported. First we have done, in table I, a comparison between the associations HWT-proposed MAP filter and DWT-soft thresholding filter.

TABLE I. A PSNRs BASED COMPARISON OF TWO DENOISING METHODS. THE FIRST ONE IS THE DENOISING METHOD PROPOSED IN THIS PAPER (PROP) AND THE SECOND IS BASED ON THE ASSOCIATION DWT-ADAPTIVE SOFT THRESHOLDING (DWT+ST).

Prop./DWT+ST	Input	Dau5	Dau7	Dau10
$\sigma_n$				
10	28.15	34.12/30.97	34.16/30.7	34.16/30.7
15	24.64	32.28/28.9	32.27/28.65	32.29/28.64
20	22.14	30.94/27.2	30.90/27.17	30.94/27.07
25	20.22	29.90/25.9	29.82/25.9	29.88/25.76
30	18.64	29.10/25.02	29.03/24.94	29.07/24.84
35	17.28	28.39/24.2	28.27/24.16	28.32/24.06

The mother wavelets used were Dau5, Dau7 and Dau10 (belonging to the Daubechies mother wavelets family and having the number indicated of vanishing moments). For the proposed MAP filter rectangular estimation windows with size 7x7 were used. The threshold was adaptively selected in the case of the second mentioned association to maximize the output peak signal to noise ratio, PSNR. Second we have compared the two types of estimation windows: rectangular and elliptic, proposed in this paper. For the left-up corner of the image Lena of size 128x128, where some directions are visible, we have obtained the results presented in table II. For any of the mother wavelets used the output PSNR obtained using directional estimation windows is superior to the corresponding value obtained using rectangular estimation windows. This conclusion remains correct for the whole Lena image. In figure 4 are represented the input image directional estimation windows is superior to the corresponding value obtained using rectangular estimation windows. This conclusion remains correct for the whole Lena image. In figure 4 are represented the input image (corresponding to an additive perturbation with white Gaussian noise having a standard deviation of  $\sigma_n=10$ ) and the denoising results obtained using the two variants of estimation windows proposed in this paper and the mother wavelets Dau10. The value of the input PSNR is of 28.16 dB (figure 4, first image) and the values of the output PSNRs are of 32.52 dB for the variant corresponding to rectangular estimation windows (figure 4, second image) and of 34.09 dB for the variant corresponding to directional estimation windows (figure 4, third image). The superiority of the method proposed in this paper is obvious. In fact these results are comparable (slightly better) with the results obtained using another WT with enhanced directionality, the contourlet transform, reported in [27]. These results can be also compared with those reported in [28], where the filtering in the HWT domain is realized

TABLE II. A COMPARISON OF TWO VARIANTS OF THE PROPOSED DENOISING METHOD OBTAINED USING RECTANGULAR (RECT) AND DIRECTIONAL (DIR) ESTIMATION WINDOWS.

Rect/direct	Input	Dau5	Dau7	Dau10
$\sigma_n$				
10	28.16	32.65/35.19	33.01/35.32	32.63/35.30
15	24.65	28.32/33.30	29.80/33.24	28.31/33.15
20	22.15	27.41/31.70	25.80/31.67	26.60/31.57
25	20.21	25.52/30.09	25.15/30.31	25.75/30.19
30	18.62	24.31/28.79	24.31/29.14	22.24/28.94
35	17.27	23.40/27.77	22.98/27.79	22.49/27.78



Figure 4. A comparison of the results obtained using the two variants of the proposed denoising method. The input image is represented in the first figure (up), the result obtained using rectangular windows is represented in the second figure (middle) and the result obtained using the directional windows is represented in the third image (bottom).

without MAP systems. Comparing the second and third images in figure 4 we can observe the superiority of the directional treatment (see for example the Lena's nose).

## V. CONCLUSION

The HWT is a very modern WT as it has been formalized only one year ago [20]. It was already used in image denoising, [28], the method obtained being named hyperanalytic denoising. In this paper we have used a very simple implementation of this transform, which permits the exploitation of the mathematical results and of the algorithms previously obtained in the evolution of wavelets theory. It does not require the construction of any special wavelet filter. It has a very flexible structure, as we can use any orthogonal or biorthogonal real mother wavelets for the computation of the HWT. Our simulation results are similar or better than the results obtained using other denoising methods based on new wavelet transforms, like for example the contourlet transform, [27]. The approach presented in this paper is different in comparison with the denoising strategy proposed in [28] from two points of view: the implementation of the 2D HWT and the filter used in the wavelets domain. We have preferred a MAP filter based on a very simple model of the wavelet coefficients of the useful image. We have compared two implementations of the MAP filter differing by the form and orientation of the windows used to estimate the local standard deviation of the useful image and we highlighted the superiority of directional windows.

The simulation results presented in this paper illustrate the effectiveness of the proposed algorithm. The comparisons made suggest the new denoising results are competitive with the best wavelet-based results reported in literature, despite the imprecision of the statistical model used. So, the less precise statistical model is not an important handicap. Of course, one of our future research directions will be the association of the new implementation of the HWT with a better MAP filter (based on a more precise model of wavelet coefficients). We intend also to optimize the directional selectivity of the proposed denoising method, by including the preferential directions 0 and  $\pi/2$  and by varying the size of directional estimation windows in function of the current resolution of the WT.

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